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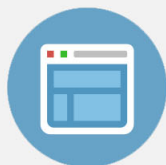
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Tunable magnetolectric response of dimensionally gradient laminate composites

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A magnetolectric (ME) sensor exhibiting wideband behavior as a function of applied magnetic DC bias and frequency was designed by combining the dimensionally gradient piezoelectric layer with Metglas magnetostrictive layers in laminate configuration. The ME coefficient of the band in the DC magnetic range of 52–242 Oe was measured to be 3000 mV/cmOe under the resonant condition of $f = 107$ kHz. The wideband in the AC magnetic field frequency range of 41–110 kHz had the ME coefficient in the vicinity of 260 mV/cmOe under the conditions of $H_{AC} = 1$ Oe and $H_{DC} = 70$ Oe. This frequency-dependent ME behavior clearly showed two different states on each side of the resonance peak which could open the possibility of developing new applications such as magnetic field-controlled switches.

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Composites with high magnitude of magnetolectric (ME) coefficient have been projected to find applications in sensors, transducers, actuators, resonators, harvesters, and servomechanism.^{1–6} It is well-known that electromechanical resonance (EMR) in ME composites can be utilized to improve the magnitude of ME coefficient.^{7–10} The EMR range can be tuned by designing the piezoelectric layers with gradient geometry.^{9,10} Gradient laminate composites in the vicinity of EMR can provide wideband behavior with high ME coefficient. Building upon this premise, we designed a ME resonator that exhibits flat and high ME response over wide magnetic DC bias and frequency range. In this study, we demonstrate that this design can provide tunable states as a function of frequency.

Dimensionally gradient piezoelectric plate⁹ with thickness of 1 mm was fabricated by mechanical polishing and dicing technique, as shown in Fig. 1(a). Piezoelectric plates with composition $Pb(Zn_{1/3}Nb_{2/3})_{0.2}(Zr_{0.5}Ti_{0.5})_{0.8}O_3$ (PZNT) were synthesized by conventional mixed oxide method.¹¹ Piezoelectric constant of poled PZNT plates was found to be 500 pC/N, and the dielectric constant was 2219 at 1 kHz. The piezoelectric voltage constant (g_{33}) was of the order of 23.41×10^{-3} Vm/N. On this PZNT plate, 25 μ m-thick Metglas (2605SA1, Metglas Inc, USA) sheets of desired dimensions were attached using epoxy (West System, USA) with the curing temperature of 80 °C. Finite element modeling software ATILA was used to model the structure and obtain resonance modes.^{9,12} Impedance spectrum of the composites was measured by LCR meter (HP 4194 A). For ME voltage coefficient measurement in longitudinal–transversal (L–T) mode configuration, an electromagnet was used to apply the DC magnetic field, and the samples were placed in the center of the Helmholtz coil under an AC magnetic field (H_{ac}). The voltage induced on the laminate was monitored by using a

lock-in amplifier. The magnetostriction was evaluated by using the strain gauge and Wheatstone bridge.

Impedance spectrum of asymmetric H -shaped piezoelectric plate was measured to identify the EMR range. In conjunction, ATILA simulation was conducted to observe the displacement behavior of H -shaped piezoelectric plate as shown in Fig. 1(b). From simulation, the first, second, and third resonances were computed to be 42, 44, and 94.3 kHz, respectively. In comparison, measured impedance spectrums for the asymmetric H -shaped piezoelectric plate exhibited resonance peaks at 42, 44, and 96 kHz, as shown in Fig. 1(b). Thus, the results between simulation and measurement were in good agreement. The first mode at 42 kHz was associated with biaxial bending of the large rectangular section and the intermediate rectangular section, while the second mode at 44 kHz was related to the biaxial bending of the large rectangular section and the small rectangular section. The third mode at 99 kHz came from the lateral displacement of the large rectangular section, as shown in Fig. 1(c).

Figure 2(a) shows the fabricated ME laminate composite based on Fig. 1(a). On top of the PZNT layer, four layers of Metglas with the area of 13×7 mm² were attached at the middle rectangular section (Section A), thirty layers of Metglas with area of 20×7 mm² were attached at the larger rectangular section (Section B), and five layers of Metglas with area of 7×7 mm² were attached at the intermediate rectangular section (Section C), as shown in Fig. 2(a). There are two variables which could be adjusted to achieve a wideband ME response.^{9,10} First, if the rectangular area of two sections in piezoelectric plate is different than the one with smaller number of Metglas layers will show higher ME coefficient. Second, if the numbers of Metglas layers are same, the rectangular section with the smaller area will show smaller ME coefficient. Thus, by adjusting the ratio of Metglas layers on various rectangular sections, a composite ME response with flat behavior can be obtained.

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Figure 2(b) shows the magnetostriction (S_{ij}) and piezomagnetic (q_{ij}) coefficient for varying dimensions and stack configurations of Metglas. In this figure, S_{11} corresponds to longitudinal in-plane magnetostriction parallel to H_{DC} and q_{11} is the longitudinal in-plane piezomagnetic coefficient corresponding to the differential of S_{11} . The maximum in-plane magnetostriction was found to be 28 ppm regardless of Metglas stack configurations and dimensions; however, the strain behavior was strongly dependent on the stack configuration and dimensions. Four layers of Metglas with the area of $13 \times 7 \text{ mm}^2$, thirty layers of Metglas with area of $20 \times 7 \text{ mm}^2$, and five layers of Metglas with area of $7 \times 7 \text{ mm}^2$ showed piezomagnetic coefficient corresponding to 0.38, 0.21, and 0.16 ppm/Oe at 70, 150, and 204 Oe of H_{DC} , respectively. These piezomagnetic behaviors will result in strong elastic coupling with the piezoelectric sections.

Figure 2(c) shows the measured ME response from the composite structure as a function of magnetic DC bias under the condition of $H_{AC} = 1 \text{ Oe}$ at $f = 1 \text{ kHz}$. The peak at 70 Oe was associated with Section A. The second peak of the ME

coefficient at 150 Oe was associated with Section C. The third peak of the ME coefficient at 209 Oe was associated with Section B. The value of DC bias at the peak in piezomagnetic coefficient corresponds to that for the peak in ME coefficient. Further, it was found that not only piezomagnetic coefficient but also piezoelectric dimensions affected elastic coupling between the Metglas and piezoelectric sections compared to Figs. 2(b) and 2(c). Furthermore, if only Section A and Section B were combined, there will be a valley in the intermediate range. By inserting Section C, the formation of flat ME band was facilitated. The overall ME response as a function of magnetic DC bias was composite of three individual responses.

The impedance and phase angle spectrums for this laminate are shown in Fig. 3(a). After Metglas was attached on the PZNT plate, multiple resonances occurred at 14, 28, 41, 44, and 107 kHz as marked with arrows in Fig. 3(a). Modal analysis conducted using ATILA is shown in Fig. 3(b). Resonant modes at 14 and 28 kHz were found to be related to the bending vibrations of all the sections. The middle section dominated the vibration mode at 14 kHz while the small

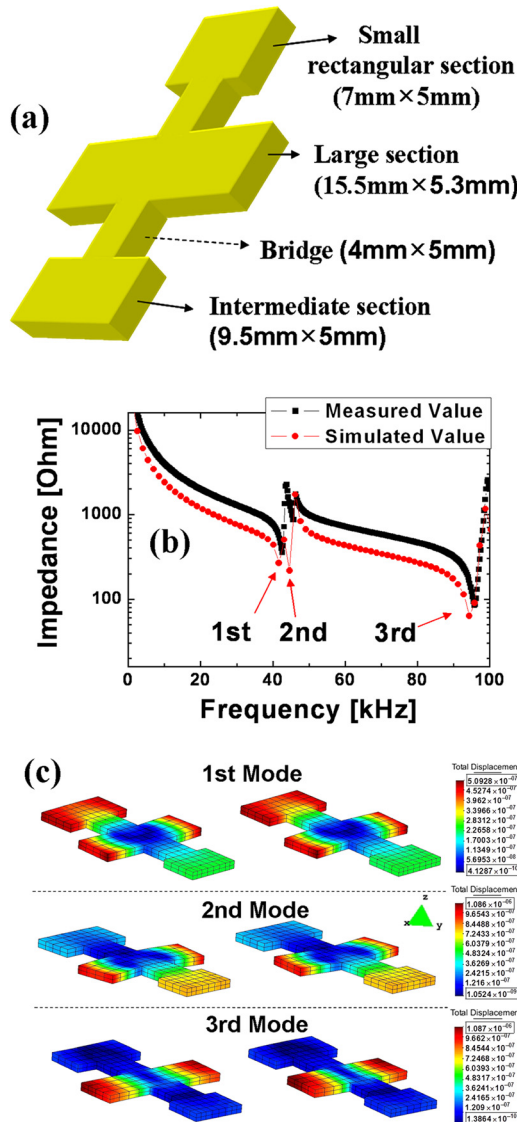


FIG. 1. (a) Schematic diagram of symmetric H-shaped piezoelectric plate, (b) impedance spectrums between simulated and measured results, and (c) FEM analysis of resonant motions at 42, 44, and 94.3 kHz.

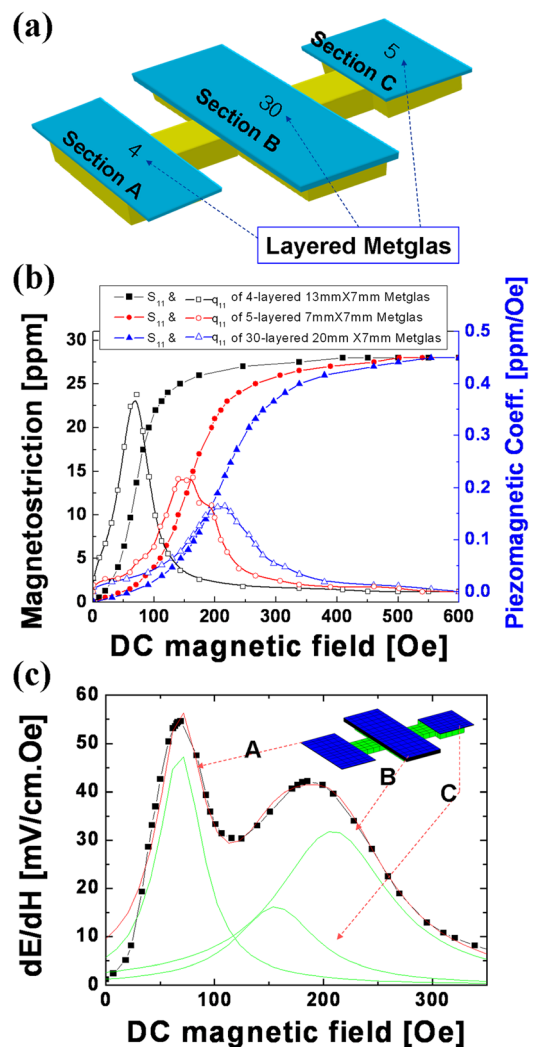


FIG. 2. (a) Schematic diagram of laminated ME composite, (b) magnetostriction (S_{11}) and piezomagnetic coefficient (q_{11}) for varying dimensions and stack configurations of Metglas, (c) ME response as a function of DC magnetic field under the constant condition of $H_{ac} = 1 \text{ Oe}$ at $f = 1 \text{ kHz}$.

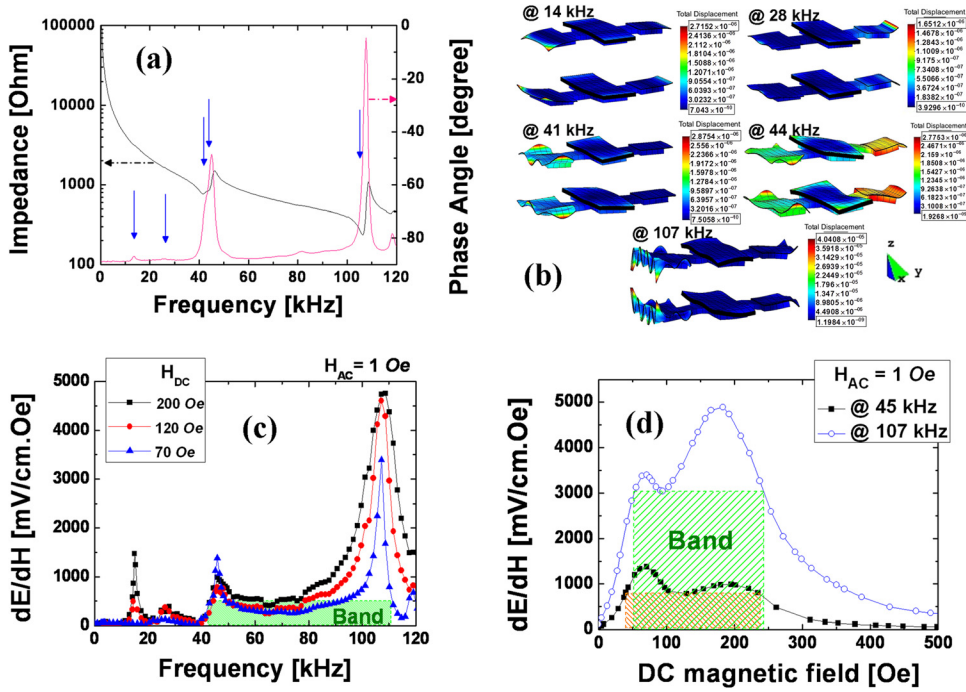


FIG. 3. (a) Impedance and phase angle spectrums of the ME composite, (b) resonant displacements at 14, 28, 41, 44, and 107 kHz determined from simulations, (c) ME response as a function of frequency, (d) ME response as a function of DC magnetic field under resonance conditions at 45 and 107 kHz.

rectangular section dominated the vibration mode at 28 kHz. The resonance at 41 kHz was found to be combination of the bending motions of large rectangular section (Section B) and middle rectangular section (Section A). The resonance at 44 kHz was related to combined bending response from the bridge and small rectangular section (Section C). The resonant modes of ME composite at 41 and 44 kHz in Fig. 3(b) were correlated to the 1st and 2nd resonance modes of PZNT layer as determined in Fig. 1(c). The resonant mode at 107 kHz had a similar vibration mode as that at 44 kHz.

The ME output voltage from composite was measured as a function of frequency under constant H_{AC} = 1 Oe but varying H_{DC} = 70, 120, and 200 Oe, as shown in Fig. 3(c). Interestingly, ME response as a function of frequency had similar behavior as that of phase angle shift shown in Figs. 3(a) and 3(c). The ME peaks are shown at 14, 28, 45, and 107 kHz corresponding to the resonance frequencies observed in the impedance measurement. The peak ME response at 45 kHz was associated with resonances at 41 and 44 kHz. The maximum ME coefficient was found to be 4740 mV/cm Oe at f = 107 kHz under H_{DC} = 200 Oe. Interestingly, the bands were successfully formed in the range of 41–110 kHz. The composite showed high ME coefficient of 260 mV/cm Oe after the resonance peak of 41 kHz, regardless of the applied DC magnetic field. These widely extended bands were derived from the combination of the resonances at 41, 44, and 107 kHz.

The ME behavior at 45 and 107 kHz as a function of DC magnetic field is shown in Fig. 3(d). The maximum ME coefficients of 1400 and 4740 mV/cm Oe was found at 45 and 107 kHz, respectively. The band at 107 kHz exhibited ME coefficient higher than 3000 mV/cm Oe from 52 to 242 Oe of H_{DC} while the band at 45 kHz exhibited ME coefficient higher than 780 mV/cm Oe from 40 to 230 Oe of H_{DC}. The bands were obtained regardless of applied DC and AC magnetic fields. The wideband was formed in both conditions of the frequency range of 41–110 kHz and DC mag-

netic ranges of 40–230 Oe and 52–242 Oe at f = 45 and 107 kHz, as shown in Figs. 3(c) and 3(d).

Interestingly, there are two noticeable states in the ME frequency response in Fig. 3(c). Under the constant DC magnetic field condition of 70 Oe, the composite showed flat ME responses in both frequency ranges of 1–11 kHz (State I) and 41–110 kHz (State II) as shown in Fig. 4. This figure also shows the schematic of resulting ME response as a function of frequency and the applied DC magnetic field conditions. State I shows a band lower than 129 mV/cm Oe before f = 41 kHz while State II shows a band higher than 260 mV/cm Oe after f = 41 kHz under the constant condition of H_{DC} = 70 Oe. The ME coefficient of State II was 2× higher than that of State I. Consequently, ΔME (=ME_{StateII} - ME_{StateI}) was 131 mV/cm Oe under the constant H_{DC} = 70 Oe. These states were found to be adjustable with changing H_{DC} conditions. The maximum value of the ME_{StateI} can be reduced by decreasing the DC magnetic field and ranges between 0 and 129 mV/cm Oe. On the other hand, the minimum value of the

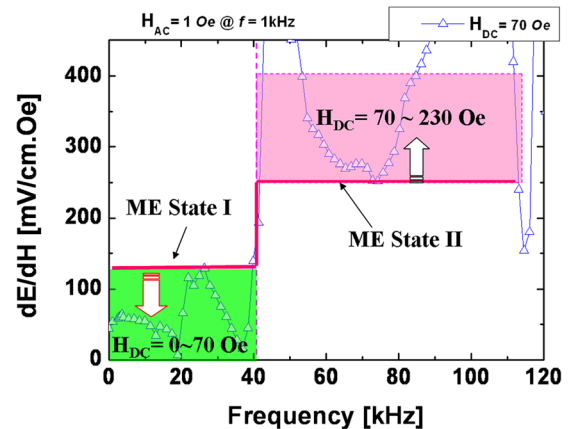


FIG. 4. Schematic of switchable ME response dependency on the frequency and DC magnetic field conditions.

ME_{StateII} can be elevated by increasing DC magnetic field condition in the range of 260–406 mV/cm Oe. Thus by tuning the magnitude of frequency and H_{DC}, ΔME in the range of 131–406 mV/cm Oe can be achieved which clearly shows the tunability of this device. These signals are strong enough to allow two distinguishable states. The clear sensing margin was seen between State I and State II: the ME values in State II were at least two times higher than that in State I. This interesting and promising ME behavior can be exploited in several applications. The wideband behavior can be a candidate for magnetic field controlled switches as well as ME harvesters. In the case of the magnetic field-controlled switches, the resonant frequency becomes the cut-off condition and the State I and State II can be considered as “off” and “on,” respectively.

Next, we attempt to explain the reason for the formation of switching states by calculating the in-plane strain and stress components for all sections and at the end combining them together to find the overall solution for mode formation. The total vibration spectrum of the laminate composite consists of several contributions. The first mode is associated with simultaneous bending vibrations of the small rectangular section with its bridge and the middle rectangular section.¹³ The equation of bending motion of *i*th-area (1 and 2 areas correspond to bridge and section) has the form:

$$\nabla^2 \nabla^2 w_i + \frac{\rho_i t_i}{D_i} \frac{\partial^2 w_i}{\partial t^2} = 0, \quad (1)$$

where $\nabla^2 \nabla^2$ is biharmonic operator, w_i is the displacement in *z*-direction, t_i is thickness, ρ_i is average density of *i*-area, and D_i is cylindrical stiffness. The strain's components can be expressed in terms of displacement as $S_{1i} = -z \frac{\partial^2 w_i}{\partial x^2}$ and $S_{2i} = -z \frac{\partial^2 w_i}{\partial y^2}$. The stress components can be expressed in terms of strains as

$$\begin{aligned} ({}^p S_k)_i &= {}^p S_{kj} ({}^p T_j)_i + {}^p d_{31} {}^p d_{3i}, \\ ({}^m S_k)_i &= {}^m S_{kj} ({}^m T_j)_i + ({}^m g_{k1})_i {}^m S_{1i}, \end{aligned} \quad (2)$$

where S_{1i} and T_{1i} are strain and stress components for *i*-area, E_{3i} is the component of electric field, H_{1i} is the component of magnetic field, s_{kj} is compliance at constant electric field for piezoelectric and at constant magnetic induction for magnetic component, g_{k1} and d_{31} piezomagnetic and piezoelectric coefficients, correspondingly. The superscripts *p* and *m* correspond to piezoelectric and piezomagnetic layers. Solving Eq. (1) for displacement of each section by using the boundary conditions given below provides the dynamic solution

$$w_1 = 0 \text{ and } \partial w_1 / \partial x = 0 \text{ at } x = 0,$$

$w_1 = w_2$, $\partial w_1 / \partial x = \partial w_2 / \partial x$, $(M_1)_1 = (M_1)_2$, and $(V_1)_1 = (V_1)_2$ on the boundary of 1 and 2 areas;

$(M_1)_2 = 0$ and $(V_1)_2 = 0$ at $x = L$ (L is the total length of section and bridge);

$(M_2)_i = 0$ and $(V_2)_i = 0$ at $y = \pm b_i/2$ (b_i is the width of *i*-area),

where $(M_j)_i$ is the moment of rotation and $(V_j)_i$ is the transverse force with respect to *j*-axis.

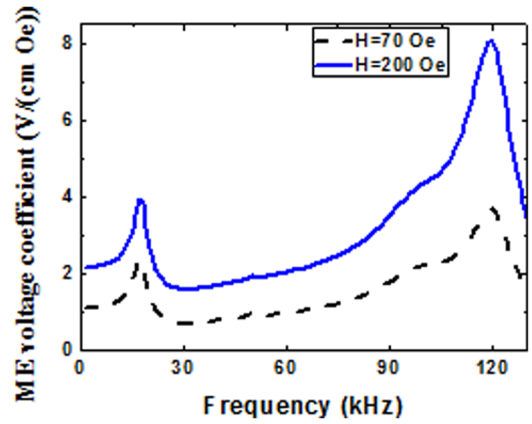


FIG. 5. Frequency dependence of ME voltage coefficient for Metglas-PZNT composite for bias field of 200 and 70 Oe.

The computed displacements were used to determine the strain components and then the stress components from Eq. (2). Substituting the stress components into open circuit condition enables the calculation of the ME voltage coefficient by taking into account condition that average electric field induced across the piezoelectric layer is estimated as integral of internal electric field taken over the piezoelectric thickness, $E = \frac{1}{t_p} \int_{t_p} {}^p E_3 dz$. These calculations were carried out numerically. The resonance frequencies for the small rectangular section and the middle rectangular section were found to be approximately equal. The second harmonic of this mode can be seen at 97 kHz. It should be noted that theoretical estimate for the next resonance frequency (at 44 kHz) as the bending mode of large rectangular section was made using similar procedure. The bridge was not included into vibrating system since its displacement (in *z* direction) was negligible. Finally, the mode at 120 kHz was calculated under assumption that it came from the axial mode of the large rectangular section. The estimate for this case was found using the computing method known heretofore.⁴ Figure 5 clearly reveals that frequency band of 90–130 kHz arises from the combination of the resonance modes. These states are adjustable with H_{DC}.

In summary, we demonstrate a ME resonator exhibiting wideband behavior by fabricating a dimensionally gradient structure and combining with laminate configuration. We were able to obtain a flat ME response in the DC magnetic bias range of 52–242 Oe where the ME coefficient was higher than 3000 mV/cm Oe under resonant condition. The promising wideband behavior as a function of frequency occurred in the range of 41–110 kHz where the ME coefficient was higher than 260 mV/cm Oe independent of applied H_{DC}. Under low H_{DC} condition, two different ME states were clearly obtained.

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