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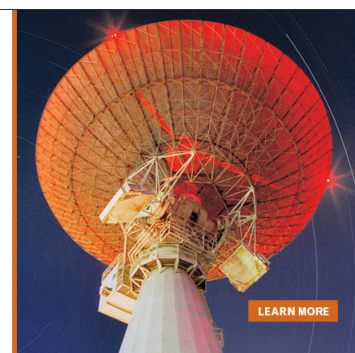
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Structural dependence of nonlinear magnetoelectric effect for magnetic field detection by frequency modulation

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The structure differences of magnetoelectric (ME) laminates for passive and active mode sensors are discussed. The Fourier coefficient A_1 calculated from the data of $\alpha_{\text{ME}}H_{\text{dc}}$ indicates that $N = 1$ (where N is the number of Metglas layers) should be the optimum structure for the active mode. Experimental investigations of the magnetic field sensitivity agree well with this conjecture. For $N = 1$, the magnetic field sensitivity was $0.66 \text{ nT/Hz}^{0.5}$, which was 3.1 times larger than for $N = 5$.
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The magnetoelectric (ME) effect—the induction of magnetization by an applied electric field (E) or polarization by magnetic field (H)—has been of recent research interests due to many potential applications.^{1–3} The ME effect in laminate composites is known to be much larger than in single phase and particulate composites, due to a combination of the magnetostrictive and piezoelectric effects of the individual layers.^{4–6} Application of magnetic field to the ME laminates produces an elastic strain in the magnetostrictive phase that is then stress coupled to that of the piezoelectric layer, resulting in an induced voltage. Significantly, higher values of the ME voltage coefficient (α_{ME}) have been reported in tri-layer Metglas/piezo-fibers/Metglas structures with a multi-push-pull configuration that is longitudinally poled and longitudinally magnetized (L-L), promising applications such as magnetic field sensors and gradiometers.⁷

In previous reports, motivated by passive magnetic sensor applications, a linear response of the ME composites to a weak AC magnetic field H_{AC} was shown.⁸ A DC magnetic bias field was applied to obtain the maximum piezomagnetic coefficient $d_{33,\text{m}}$, yielding the highest induced ME voltage (V_{ME}) to H_{AC} . Recently, ME laminate composites have also been used in an active mode to sense small DC magnetic fields or weak AC magnetic fields, via a frequency modulation technique.^{9,10} Such an approach has been utilized to reduce the environmental vibrational noise and $1/f$ noise in the low frequency range. In a magnetically unshielded environment, the sensitivity to small changes in H_{AC} can be enhanced by at least two orders magnitude compared to passive magnetic sensors.¹⁰ However, to date, all investigations of the nonlinear ME coefficient, which is the physical basis of the frequency modulation method, have been reported using the same laminate structure as that used in the passive mode.¹¹ Due to a dependence of the different orders of nonlinearity on the magnetostrictive material, the structure resulting in a maximum nonlinear ME coefficient may not be the same as for the linear ME effect used in the passive mode. However, any investigations of the ME laminate structure that optimize the mode have not yet been performed.

In this letter, the structure differences of ME laminates for sensor applications in passive and active modes are discussed. Theoretical predictions and experimental results both illustrate that $N = 1$ is the optimum structure for the active ME sensors. For $N = 1$, the magnetic field sensitivity was 3.1 times larger than for $N = 5$.

Several Metglas/PZT/Metglas multi-push-pull L-L mode laminates were made.¹² A $40 \text{ mm} \times 10 \text{ mm}$ PZT bundle served as the core, which consisted of five $40 \text{ mm} \times 2 \text{ mm}$ PZT-5A fibers (Smart Materials, Sarasota, FL) oriented along the length direction of the laminates. Two interdigitated (ID) Kapton electrodes were bonded to the top and bottom surfaces of the piezoelectric bundle with an epoxy resin (Stycast 1264, USA). The width and the separation of the electrodes were chosen to be 0.15 mm and 1 mm , respectively. Different numbers N of Metglas layers (Vitrovac 7600F, Hanau, German) of dimensions $80 \text{ mm} \times 10 \text{ mm}$ were bonded to both sides of the PZT core composite for $N = 1$ to 7.

The value of α_{ME} was then measured using a lock-in amplifier (SR-850) in response to a pair of Helmholtz coils driven at an AC magnetic field of $H_{\text{AC}} = 0.1 \text{ Oe}$ at a frequency of $f = 1 \text{ kHz}$. A DC magnetic field was applied by a large electromagnet. Figure 1(a) shows the values α_{ME} as a function of H_{DC} for Metglas/PZT/Metglas laminates with different N . It can be seen that α_{ME} increased as H_{DC} was increased; and subsequently decreased as H_{DC} was further increased. The maximum value of α_{ME} increased with increasing number of Metglas layers until $N = 5$, and then decreased with further increase in N . For the optimum value of N , the magnetic field sensitivity of the passive ME sensor was about $20 \text{ pT/Hz}^{0.5}$.^{13,14}

For a constant applied AC magnetic field, the output voltage V_{ME} is proportional to α_{ME} . The value of V_{ME} can be then given as a five-order Fourier expansion series,¹⁵ as

$$V_{\text{ME}} = a_0 + a_1H + a_2H^2 + a_3H^3 + a_4H^4 + a_5H^5, \quad (1)$$

where $H = H_{\text{mod}} \cos(2\pi f_{\text{mod}}t)$ is an applied AC magnetic field, H_{mod} is the modulation amplitude, and a_i ($i = 0, 1, 2, \dots$) are the i^{th} order Fourier coefficients of the field. The values of a_i can be estimated by fitting (1) to the $\alpha_{\text{ME}}H_{\text{dc}}$ data of Fig. 1(a), as shown in the Appendix for $N = 1$ to 7. At the

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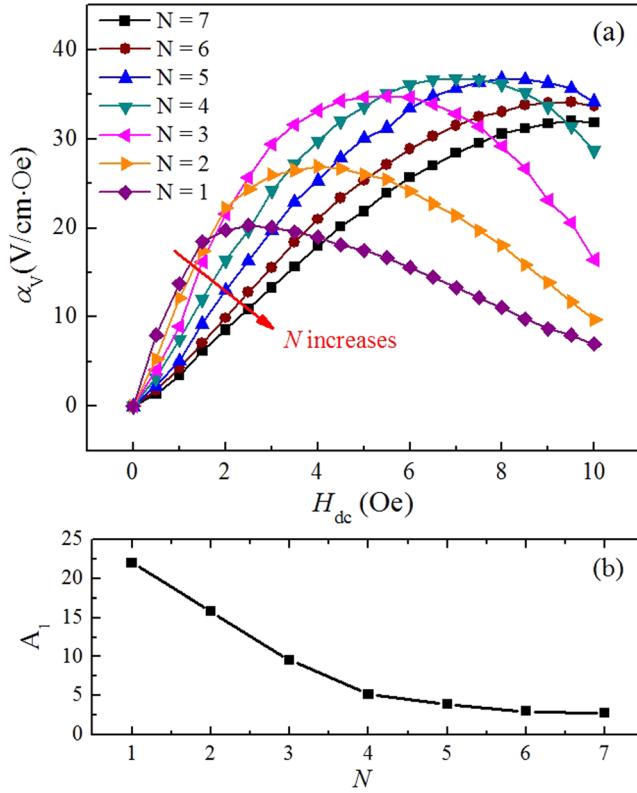


FIG. 1. (a) α_{ME} as a function of H_{dc} for Metglas/PZT/Metglas laminates with different values of N , and (b) values of A_1 calculated by Eq. (2) for various N .

modulation frequency f_{mod} , V_{ME} is proportional to a rearranged Fourier coefficient A_1 , given as

$$A_1 = a_1 H_{mod} + \frac{3}{4} a_3 H_{mod}^3 + \frac{5}{8} a_5 H_{mod}^5. \quad (2)$$

Figure 1(b) shows the calculated values of A_1 for various N . With increasing N , the value of A_1 decreases, implying for $N=1$ that V_{ME} is the highest. These results illustrate that the optimum structure of Metglas/PZT/Metglas laminate for the active mode is different than that for the passive one.

Next, the M - H hysteresis loops were measured by a vibrating sample magnetometer (VSM) for various values of N , as shown in Figure 2(a). The value of the magnetic susceptibility χ was then calculated as dM_s/dH , where M_s is the saturation magnetization. In Figure 2(b), the value of χ can be seen to be maximum for $N=1$, and to decrease as N increases. This decrease was due to the constraint imposed by epoxy bonding layers between the Metglas layers. As N is increased, the additional epoxy layers increasingly constrain the magnetostriction of the Metglas layers. Such constraint makes the longitudinal magnetization more difficult.¹⁶ To achieve an optimum magnetostriction, the value of the required H_{DC} has to be increased as N is increased. Thus, χ effectively decreases with increasing N .¹⁷ For an AC modulation field H_{mod} and a small signal field H_{AC} , the magnetostriction λ of the Metglas layers at the frequency of $f_{mod} - f_{AC}$ can then be given as¹⁸

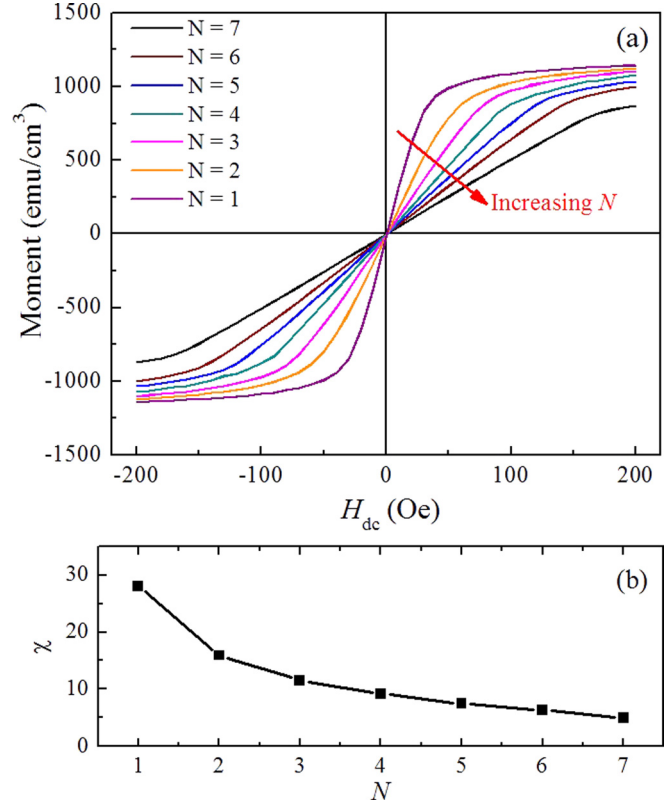


FIG. 2. (a) M - H hysteresis loops and (b) χ for various values of N .

$$\lambda = \frac{3\lambda_s \chi^2}{2M_s^2} H_{mod} H_{AC}, \quad (3)$$

where λ_s is the saturation magnetostriction. From (3), it can be seen that V_{ME} is directly proportional to χ^2 . Thus, to get the highest value of the output signal, the structure that yields the maximum value of χ should be chosen, which is $N=1$.

Finally, an incident magnetic field of $H_{mod} = 1$ Oe at $f_{mod} = 1$ kHz was applied using a Helmholtz coil and an AC field of $B_{AC} = 70$ nT at a $f_{AC} = 1$ Hz using a drive coil. The modulation field H_{mod} was then measured using a SR-785 dynamic signal analyzer (Stanford Research Systems). Figure 3(a) shows the measured modulation spectrum in the frequency range between 998.5 Hz and 1001.5 Hz for various N . As shown in Figure 3(b), for $N=1$, the ME output voltage V_{ME} at $f_{mod} - f_{AC}$ was maximum; and subsequently decreased with increasing N . At $f = f_{mod} - f_{AC}$, the voltage noise V_{noise} for $B_{AC} = 0$ at the value of each N was also measured as shown on the right axis of Fig. 3(b). With increasing N , V_{noise} also increased slightly, due to an increased value of α_{ME} at low fields (i.e., $H_{DC} < 1$ Oe). The magnetic field sensitivity can be estimated as $V_{ME}/B_{AC} * V_{noise}$, as shown in Figure 3(c). With increasing N , the active sensor had an enhanced magnetic field sensitivity. The magnetic field sensitivity was 0.66 nT/Hz^{0.5} for $N=1$, which was 4.2 times larger than for $N=7$ and 3.1 times larger than for $N=5$. Clearly, the optimum structure for the ME active mode structure ($N=1$) is different than the passive one ($N=5$). The increased background noise by modulation mainly limits the sensitivity.^{10,15}

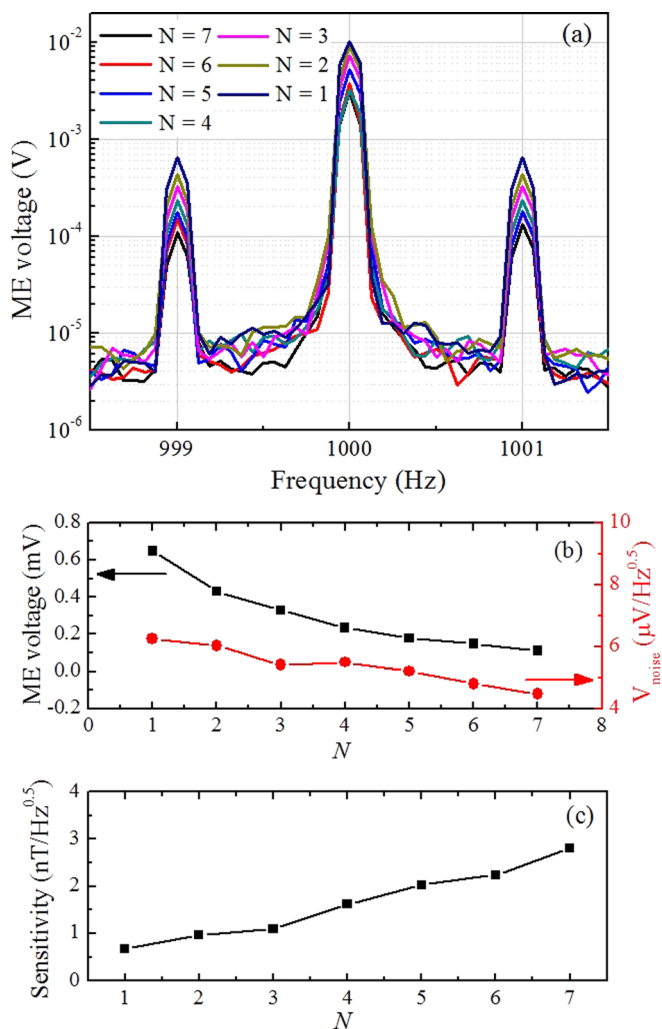


FIG. 3. (a) Measured modulation spectrum taken over the frequency of 998.5 Hz to 1001.5 Hz for various values of N ; (b) V_{ME} and V_{noise} at $f = f_{mod} - f_{AC}$; and (c) magnetic field sensitivity as a function of N .

If the undesired background noise could be reduced, the sensitivity could be further enhanced by two orders of magnitudes.

In conclusion, we revealed the structure differences of ME laminate for sensor applications in passive and active modes. In active mode, the calculated Fourier coefficient A_1 indicates that the ME output voltage is highest for $N=1$. Experimental results agree well with the prediction, and are explained by the change of χ . The magnetic field sensitivity is enhanced by optimizing structure.

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APPENDIX: ESTIMATED FOURIER COEFFICIENTS A_i FROM FIGURE 1(A)

N	1	2	3	4	5	6	7
a_0	-0.27	-0.76	-0.7	-0.06	-0.17	-0.03	-0.13
$a_1(\text{Oe}^{-1})$	21.1	15.8	10.2	5.7	4.25	3.21	2.94
$a_2(\text{Oe}^{-2})$	-7.76	-2.51	1.57	2.31	1.98	1.43	1.08
$a_3(\text{Oe}^{-3})$	1.3	0.02	-0.8	-0.71	-0.54	-0.33	-0.24
$a_4(\text{Oe}^{-4})$	-0.11	0.019	0.091	0.071	0.053	0.027	0.021
$a_5(\times 10^{-3} \text{Oe}^{-5})$	3.50	-1.10	-3.50	-2.60	-1.90	-0.896	-0.752

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